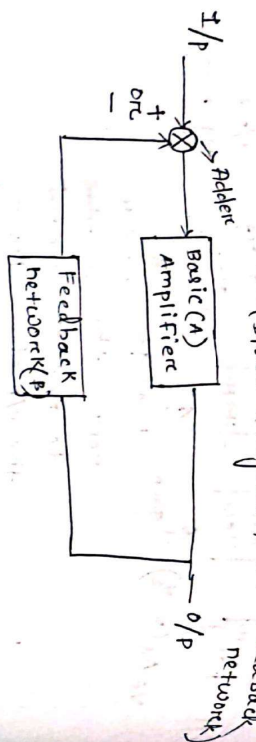


Feedback Amplifier :-

(Block diagram of feedback network)



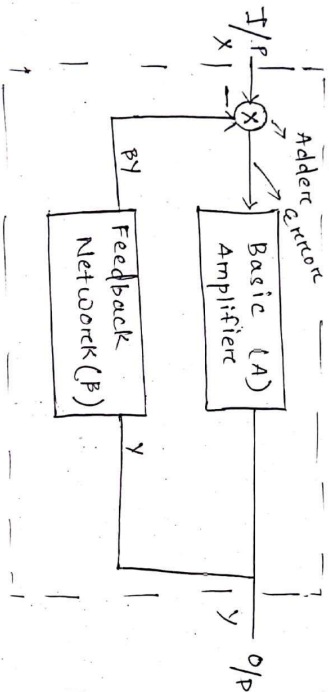
→ Feedback network is used to feedback circuit control the o/p of the system by adding or subtracting some part of o/p in the i/p

Types of feedback :-

(1) Negative Feedback :-

→ It is also known as degenerative circuit.

Ex:- All amplifiers



$$\text{error} = x - BY$$

$$Y = A \cdot \text{error}$$

$$\Rightarrow Y = A(x - BY)$$

$$\Rightarrow Y = AX - APY$$

$$\Rightarrow Y + APY = AX$$

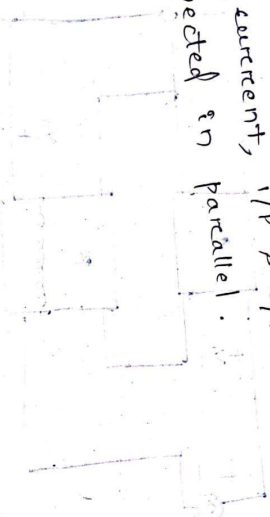
$$\Rightarrow Y(1 + AP) = AX$$

$$\Rightarrow \frac{Y}{X} = \frac{A}{1 + AB}$$

= Gain of feedback amplifier

Concept of -ve feedback:-

- All the o/p should be reached at the i/p of feedback (B) network, if the o/p is voltage the connection ~~of~~ ⁱⁿ betⁿ i/p & B network and o/p ~~connection~~ must be ⁱⁿ shunt / parallel.
- If the o/p is current, then the connection betⁿ i/p of B network and o/p must be in series.
- Input & output of B network are connected in ~~a~~ such a manner that, both will have significance.
- If the i/p is current, i/p & o/p of an B network must be connected in parallel.



- If the i/p is voltage i/p & o/p of B network must be in series.

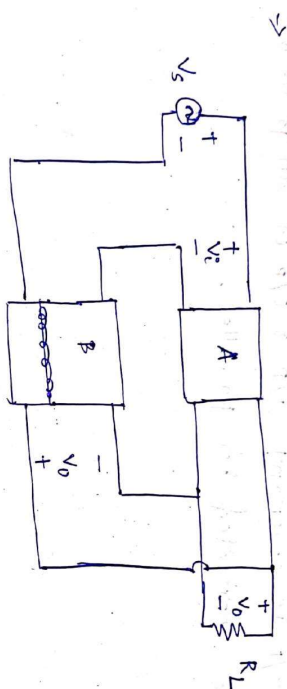
-Ve feedback connection types (types of feedback amplifiers) :- (Topology)

→ There are 4 types of feedback amplifier

- (i) Voltage amplifier
- (ii) Current amplifier
- (iii) Trans impedance
- (iv) Trans conductance.

Voltage Amplifier :-

OR Voltage-series or series-shunt.



$$V_f = B V_o$$

$$\text{Gain with feedback} = A_f \triangleq \frac{V_o}{V_s}$$

$$A_f = \frac{V_o}{V_s}$$

Applying KVL at I/P,

$$V_s - V_i - V_f = 0$$

$$\Rightarrow V_s = V_i + V_f$$

$$A = \frac{V_o}{V_i}$$

$$A_f = \frac{V_o}{V_s}$$

$$= \frac{V_o}{V_i + V_f}$$

$$= \frac{\frac{V_o}{V_i}}{\frac{V_i + V_f}{V_i}}$$

$$= \frac{A}{1 + \frac{V_f}{V_i}} = \frac{A}{1 + \frac{V_f}{V_o} \times \frac{V_o}{V_i}}$$

$$A_f = \frac{A}{1 + A\beta}$$

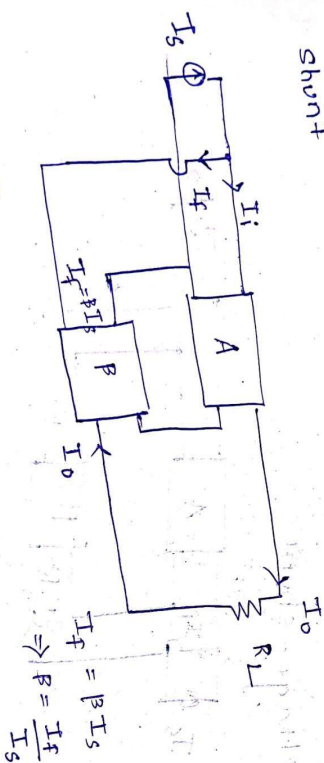
Current Amplifier:-

OR shunt-series OR current shunt

$$\frac{I_o}{I}$$

$$\frac{D/P}{I}$$

shunt



$$A_f = \frac{I_o}{I_s}$$

Applying KCL at input

$$I_s = I_i + I_f$$

$$A_f = \frac{I_o}{I_i + I_f}$$

$$A = \frac{I_o}{I_i}$$

$$A_f = \frac{I_o}{I_i}$$

$$\frac{I_i + I_f}{I_i}$$

$$= \frac{A}{1 + \frac{I_f}{I_i}}$$

$$= \frac{A}{1 + \frac{I_f}{I_o} \times \frac{I_o}{I_i}}$$

$$A_f = \frac{A}{1 + AB}$$

Trans-impedance :-

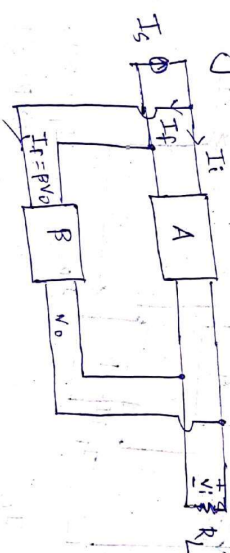
$$\frac{I/P}{I}$$

shunt

shunt

Name:- shunt-shunt

Voltage-shunt



$$I_f = P V_o \Rightarrow P = \frac{I_f}{V_o}$$

$$\text{Here, } A_f = \frac{V_o}{I_s}$$

Applying KCL at I/P

$$I_s = I_i + I_f$$

$$A_f = \frac{V_o}{I_i + I_f}$$

$$\frac{I_0}{I_i + I_f}$$

$$\frac{A}{1 + \frac{I_f}{I_i}}$$

$$\frac{A}{1 + \frac{I_f}{I_i} \cdot \frac{V_o}{V_i}}$$

$$A_f = \frac{A}{1 + A\beta}$$

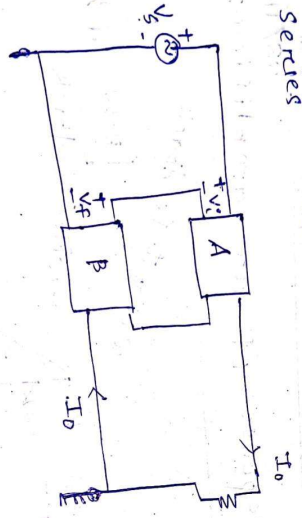
Trans-conductance :-
(series-series or current series)

$$\frac{I/P}{V}$$

$$\frac{O/P}{I}$$

Series

Series



$$A_f = \frac{I_o}{V_s}$$

Applying

KVL at I/P.

$$V_s = V_i + V_f$$

$$A_f = \frac{I_o}{V_i} = \frac{I_o}{V_i + V_f} = \frac{I_o}{V_i \left(1 + \frac{V_f}{V_i} \right)}$$

$$\Rightarrow A_f = \frac{A}{1 + \frac{V_f}{V_i}}$$

$$= \frac{A}{1 + \frac{V_f}{V_i} \cdot \frac{I_o}{I_o} \cdot \frac{I_o}{V_i}}$$

$$= \frac{A}{1 + \beta A}$$

$$\Rightarrow A_f = \frac{A}{1 + \beta A}$$

$\frac{I}{P} \text{ imp.}$

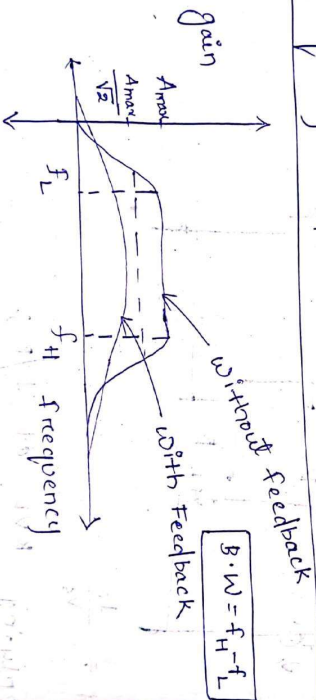
$$\star \text{ Voltage Amplifier} \rightarrow A_{Vf} = \frac{A_v}{1 + \beta A_v} \quad \uparrow Z_{if} = Z_i (1 + \beta A_v)$$

$$\star \text{ Current Amplifier} \rightarrow A_{If} = \frac{A_i}{1 + \beta A_i} \quad \downarrow Z_{if} = \frac{Z_i}{1 + \beta A_i}$$

$$\star \text{ Trans Impedance Amplifier} \rightarrow A_{zf} = \frac{A_z}{1 + \beta A_z} \quad \downarrow Z_{if} = \frac{Z_i}{1 + \beta A_z}$$

$$\star \text{ Trans conductance} \rightarrow A_{gf} = \frac{A_g}{1 + \beta A_g} \quad \uparrow Z_{if} = Z_i (1 + \beta A_g)$$

Frequency response of an amplifier:-



* Note :-

Gain $\downarrow \times$ B.W $\uparrow$ = constant

$$B.W|_f = (1 + \beta A) B.W$$

The B.W of amplifier will increase the

amplifier with -ve feedback by factor of $(1+AB)$ term.

Distortion (Noise) :- ↓

$$D_f = \frac{D}{1+AB}$$

Advantages :- Negative feedback produces the stable output by reducing the gain.

O/P imp.

$$\downarrow Z_{of} = \frac{Z_o}{1+AB}$$

$$\uparrow Z_{of} = Z_o(1+AB)$$

$$\downarrow Z_{of} = \frac{Z_o}{1+AB}$$

$$\uparrow Z_{of} = Z_o(1+AB)$$

$$A_f = \frac{A}{1+AB}$$

$A_f \rightarrow$ Gain with feedback

$A \rightarrow$ Gain without feedback

$B \rightarrow$ Feedback factor

$AB \rightarrow$ closed loop gain.

$\rightarrow$ Negative feedback enhances the o/p impedance of amplifier.

$\rightarrow$ -ve feedback reduces the o/p impedance of voltage amplifier. $Z_{if} = Z_i(1+AB)$

$$Z_{of} = \frac{Z_o}{1+AB}$$

$$Z_{of} = \frac{Z_o}{1+AB}$$

$\rightarrow$ -ve feedback enhances the B.W_f = $(1+AB)$ B.W

$\rightarrow$ -ve feedback reduces the noise (distortion).

$$D_f = \frac{D}{1+AB}$$

Q:- When -ve voltage feedback is applied to an amplifier with gain $(A) = 1000$. Overall gain falls to $A_f = 6$. Calculate the feedback factor (B) ?

Ans :-

$$A = 1000$$

$$A_f = G$$

$$\text{Here, } A_f = \frac{A}{1 + A\beta}$$

$$\Rightarrow G = \frac{1000}{1 + 1000\beta}$$

$$\Rightarrow G + 6000\beta = 1000$$

$$\Rightarrow 6000\beta = 1000 - G$$

$$\Rightarrow \beta = \frac{1000 - G}{6000} = \frac{1000 - 1000}{6000} = 0.16$$

Q:- The gain & distortion of an amplifier are 150 & 5% respectively without feedback. If the stage has 10% of its o/p voltage apply as -ve feedback, Find the distortion of amplifier with -ve feedback.

Ans :-

$$A = 150$$

$$D = 5\%$$

$$\beta = 10\% = \frac{1}{10}$$

$$D_f = \frac{D}{1 + A\beta} = \frac{5\%}{1 + 150 \times \frac{1}{10}} = \frac{5\%}{16} = 0.003$$

Q:- Determine the voltage gain, i/r & v/p impedance with feedback for a voltage series feedback having $A = -100$, $R_i = 10k\Omega$, $R_o = 20k\Omega$.

$$R_o = 20k\Omega$$

$$(a) \beta = -0.1$$

$$(b) \beta = -0.5$$

$$(a) A_f = \frac{A}{1 + A\beta} = \frac{-100}{1 + (-100)(-0.1)} = \frac{-100}{101} = -0.99$$

$$x_{if} = x_i (1 + A\beta) =$$

$$x_{of} = \frac{x_o}{1 + A\beta} =$$

Q: If an amplifier with gain $A = -1000$, with $\beta = -0.1$ has a gain change of 20% due to temperature. Calculate the change in gain of calculate feedback.

Ans:-

$$A = -1000$$

$$\beta = -0.1$$

$$1000 \times \frac{1}{10}$$

$$A_F = \frac{A}{1 + A\beta}$$

$$\frac{dA_F}{dA} = \frac{(1 + A\beta)(1 - A\beta)}{(1 + A\beta)^2} = \frac{1}{(1 + A\beta)^2}$$

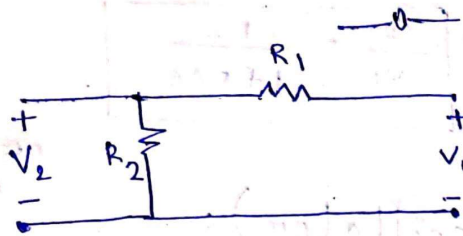
$$dA_F = \frac{dA}{(1 + A\beta)^2} = \frac{20\%}{1 + (-1000)(-0.1)} = \frac{20\%}{1 + 100}$$

$$= \frac{\frac{20}{100}}{\frac{101}{5}} = \frac{1}{505} = 0.0019$$

Wednesday

02/12/19

①



$$\frac{V_2}{V_1} = ?$$

$$\frac{1}{\frac{1}{R_1} + \frac{1}{R_2}} = \frac{R_1 R_2}{R_1 + R_2}$$

In series, voltage division occurs.

$$V_2 = \frac{V_1 R_2}{R_1 + R_2}$$

$$\Rightarrow \frac{V_2}{V_1} = \frac{R_2}{R_1 + R_2}$$

★ Impedance offered by

capacitor, $\frac{1}{j\omega C} = \frac{1}{2\pi fC}$
inductor, $= j\omega L = \text{ohm}$
resistor, $= R$.



unit = ohm

$L = 1 \text{ Henry}$

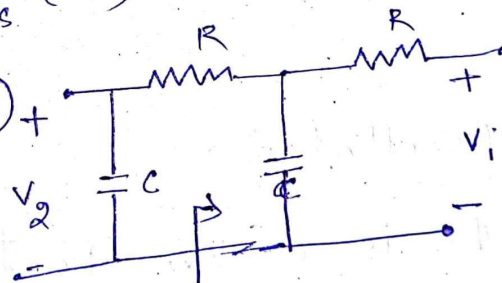
$$i, j = \sqrt{-1}$$

$$\star \bar{Z}_C = \frac{1}{j\omega C} = \frac{-1}{\omega C} = \frac{1}{Cs} (\Omega)$$

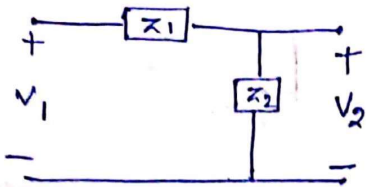
$$\star \bar{Z}_L = j\omega L = Ls (\Omega)$$

$$\star \boxed{\bar{Z} = R + jX}$$

$$\star s = \sigma + j\omega$$



★. NOTE :-



$$V_2 = V_1 \times \frac{Z_2}{Z_1 + Z_2}$$

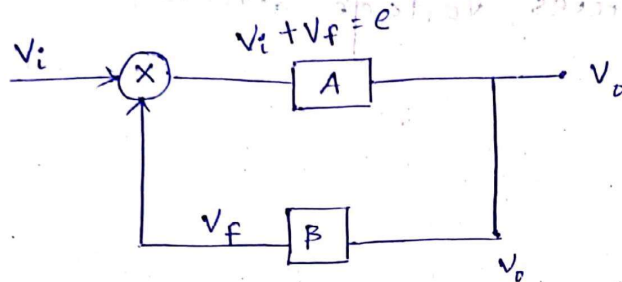
$Z = \text{Impedance}$

$$\Rightarrow V_2 = \frac{V_1 \times \frac{1}{Cs}}{R + \frac{1}{Cs}} = \frac{V_1}{Rcs + 1}$$

$$\frac{V_2}{V_1} = \frac{1}{1 + Rcs}$$

05/12/2019

Positive Feedback (Oscillator) :-



$$\beta = \frac{V_f}{V_o}$$

$e \rightarrow \text{error}$ $\frac{V_o}{e} = A$

$$V_f = \beta V_o$$

$$\Rightarrow V_o = eA = (V_i + V_f)A$$

$$= V_i A + \beta V_o A$$

$$V_o = \sin \omega_o t$$

$$V_o (1 - A\beta) = V_i A$$

$$A_f = \frac{V_o}{V_i} = \frac{A}{1 - A\beta}$$

Oscillator :-

It is a +ve feedback circuit in which there is no external input is applied and it generate and produces a sino-sodial signal (single frequency component).

Sustain Oscillator:-

When the o/p signal is a sinusoidal with one frequency and the peak amplitudes are fixed that signal is known as sustain oscillated signal.

Condition for sustain oscillation:-

(Barkhausen's criterion of principle)

$$A_f = \frac{A}{1 - AB}$$

$$1 - AB = 0$$

$$AB = 1 = 1 + j0 = 1 \angle 0^\circ \text{ or } 360^\circ$$

$$|AB| = 1, \angle AB = 360^\circ$$

$$i = j = \sqrt{-1}$$

$$|z| = \sqrt{x^2 + y^2} \quad \theta = \tan^{-1} \left| \frac{y}{x} \right|$$

$$z = x + jy$$

$$= |z| e^{j\theta}$$

$$= |z| \angle \theta$$

$$z = 1 + j1$$

$$= \sqrt{2} \angle 45^\circ$$

$$z = j = 0 + j1$$

$$= 1 \angle 90^\circ$$

$$z = 1 + j0$$

$$= 1 \angle 0^\circ \text{ or } 360^\circ$$

$\beta \rightarrow$ May be real or may be complex.

gain = Real number

i.e. $A = \text{real}$.

$\beta \rightarrow$ Feedback factor / Gain of β network.
 $A \rightarrow$ Gain of basic amplifier / Gain without feedback

$AB \rightarrow$ Loop gain.

Principles :-

1. Magnitude of loop gain = 1
 $|A\beta| = 1$ (magnitude condⁿ)

2. Phase of loop gain = 360° .
 $\angle A\beta = 360^\circ$ (Phase condⁿ)

3. $\beta = \beta_r + j\beta_i$

For sustain oscillation,

$$A\beta = 1$$

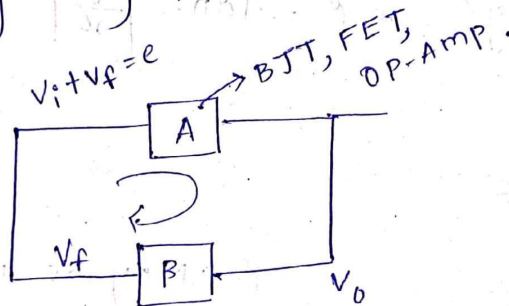
$$A(\beta_r + j\beta_i) = 1$$

$$\text{Im}(\beta) = \beta_i = 0 \text{ (It must be 0)}$$

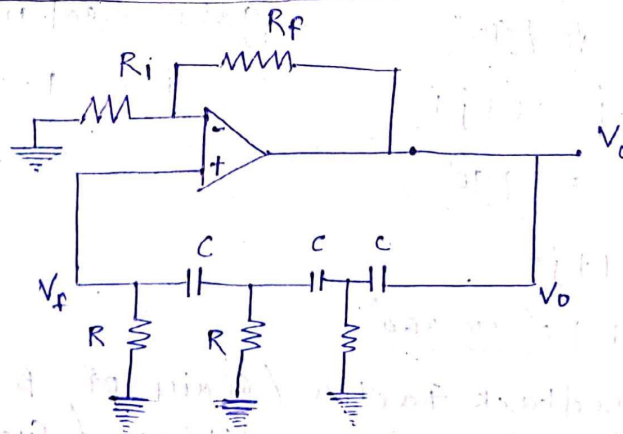
Procedure to calculate frequency of oscillation (ω_0, f_0)

→ calculate β .

→ Put imaginary part of β to calculate f_0 .



RC phase shift oscillator:-



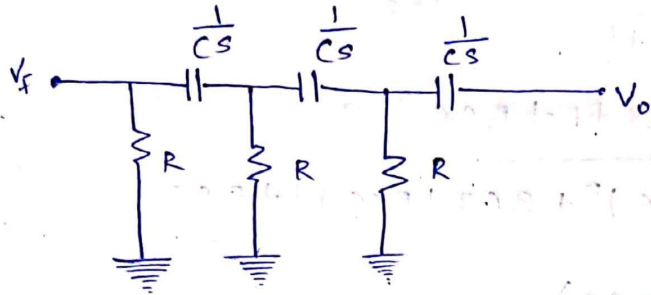
Note :-

$$Z_C = \frac{1}{j\omega C} = \frac{1}{sC}$$

$$Z_L = j\omega L = sL$$

$$Z_R = R$$

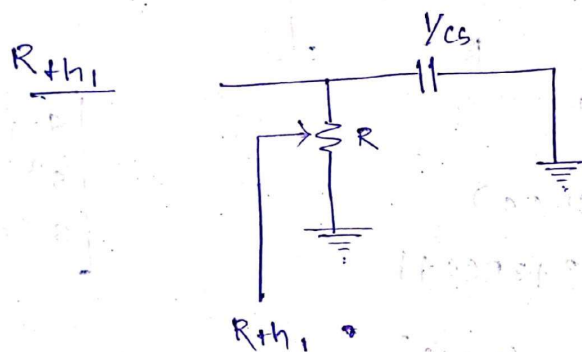
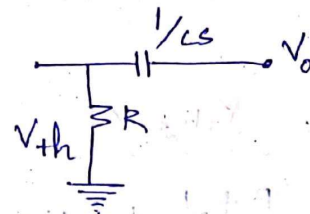
- Q.1
- | |
|----|
| 2 |
| 3 |
| 4 |
| 5 |
| 6 |
| 7 |
| 8 |
| 9 |
| 10 |
| 11 |
- 1 mark. compulsory.
- 2 mark
- Lq.



Derivation of frequency of oscillator:-

$$V_{th1} = \frac{V_o \times R}{R + \frac{1}{sC}}$$

$$= \frac{V_o R sC}{R sC + 1}$$

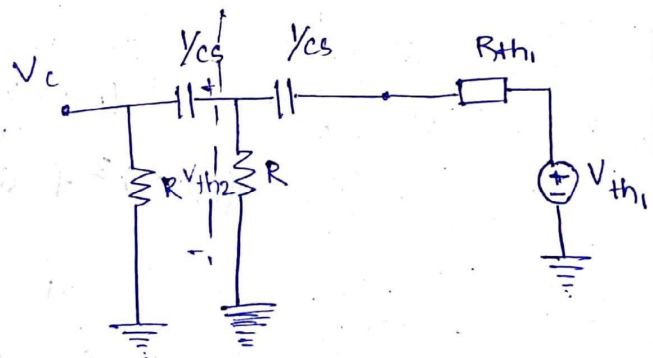


$$R_{th1} = \frac{R \times \frac{1}{sC}}{R + \frac{1}{sC}}$$

$$V_{th2} = \frac{V_{th1} R}{\frac{1}{sC} + \frac{R}{R sC + 1} + R}$$

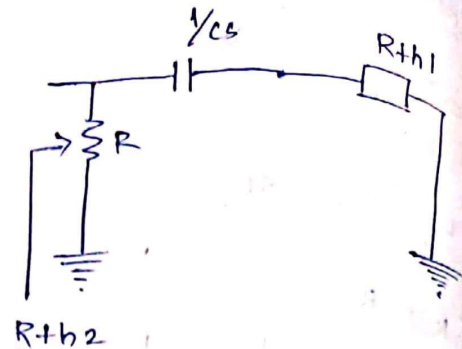
$$= \frac{V_o R sC}{1 + R sC} \cdot R$$

$$\frac{1 + R sC + R sC + (R sC)^2 + R sC}{sC (1 + R sC)}$$



$$= \frac{V_o (RCs)^2}{(RCs)^2 + 3RCs + 1}$$

$$R_{th2} = \frac{R \left(\frac{1}{Cs} + \frac{R}{RCs+1} \right)}{R + \frac{1}{Cs} + \frac{R}{RCs+1}}$$



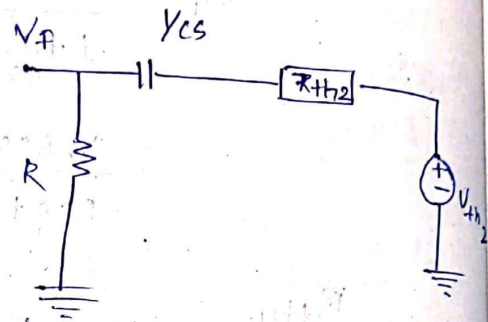
$$R_{th2} = \frac{R(RCs+1+RCs)}{(RCs)^2 + RCs + RCs + 1 + RCs}$$

$$= \frac{R(1+2RCs)}{(RCs)^2 + 3RCs + 1}$$

$$V_f = \frac{V_{th2} \times R}{R + \frac{1}{Cs} + R_{th2}}$$

$$= \frac{V_o (RCs)^2}{(RCs)^2 + 3RCs + 1} \times R$$

$$= \frac{R + \frac{1}{Cs} + \frac{R(1+2RCs)}{(RCs)^2 + 3RCs + 1}}$$



$$\begin{aligned} s &= j\omega \\ s^2 &= -\omega^2 \\ s^3 &= -j\omega^3 \end{aligned}$$

$$B = \frac{V_f}{V_o} = \frac{(RCs)^3}{(RCs)^3 + 3(RCs)^2 + RCs + (RCs)^2 + 3RCs + 1 + RCs + 2(RCs)^2}$$

$$= \frac{(RCs)^2}{(RCs)^3 + 6(RCs)^2 + 5RCs + 1}$$

$$= \frac{1}{1 + \frac{6}{RCs} + \frac{5}{(RCs)^2} + \frac{1}{(RCs)^3}}$$

$$= \frac{1}{1 - \frac{j6}{R^2 C^2 \omega^2} - \frac{5}{R^2 C^2 \omega^2} + j \frac{1}{R^3 C^3 \omega^3}}$$

$$\Rightarrow B = \frac{1}{1 - \frac{5}{R^2 C^2 \omega^2} - j \left(\frac{G}{RC\omega} - \frac{1}{R^3 C^3 \omega^3} \right)}$$

For sustain oscillation,

$$\frac{G}{RC\omega} - \frac{1}{R^3 C^3 \omega^3} = 0$$

$$\Rightarrow \frac{G}{RC\omega} = \frac{1}{R^3 C^3 \omega^3}$$

$$\Rightarrow GR^2 C^2 \omega^2 = 1$$

$$\Rightarrow \omega^2 = \frac{1}{GR^2 C^2}$$

$$\Rightarrow \omega = \frac{1}{\sqrt{G} RC}$$

$$\Rightarrow f_0 = \frac{1}{2\pi\sqrt{G} RC}$$

$$A\beta_{rc} = 1$$

$$\Rightarrow A \frac{1}{1 - \frac{5}{R^2 C^2 \omega^2}} = 1$$

$$\Rightarrow A = 1 - \frac{5}{R^2 C^2 \omega^2}$$

$$\Rightarrow A = 1 - \frac{5}{R^2 C^2 \times \frac{1}{G R^2 C^2}} = 1 - 30 = -29$$

$$\Rightarrow A = -29$$

60.